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(54) Multilevel transmission system and method with adaptive mapping

(57) A transmission system is disclosed in which a multilevel modulated signal is transmitted. The soft output information of a channel decoder is fed back and utilised by a soft demapping device in order to improve the decoding result by further iterative decoding steps. The receiver includes a demapper for generating a demapped signal, bit deinterleaver for generating a demapped and deinterleaved signal and a decoder for

generating soft reliability values representative of the decoded signal. These soft reliability values are then bit interleaved and fed back to the demapper, as *a priori* knowledge, for use in further iterations of the decoding process. Two mappings are mixed adaptively dependent on the channel conditions and the number of iterations to be used.

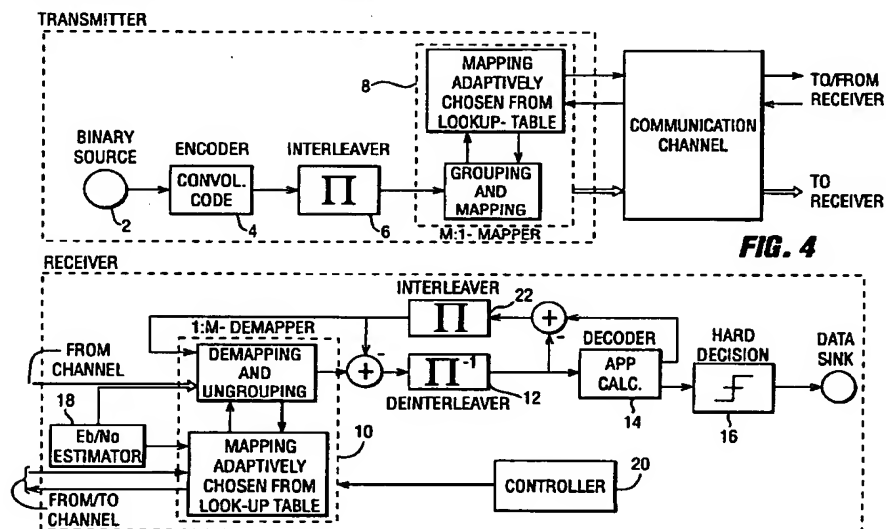


FIG. 4

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Description

[0001] This invention relates to a transmission of digital signals, e.g. in a digital radio communication system.

[0002] Iterative decoding algorithms have become a vital field of research in digital communications. The first discovered and still most popular encoding scheme suited for iterative decoding is the parallel concatenation of two recursive systematic convolutional codes, also referred to as 'Turbo Codes'. The underlying 'Turbo Principle' is applicable more generally to other algorithms used in modern digital communications, and in the past few years, other applications of the "Turbo Principle" have been found.

[0003] Channel coding is used to make the transmitted digital information signal more robust against noise. For this the information bit sequence is encoded at the transmitter by a channel encoder and decoded at the receiver by a channel decoder. In the encoder redundant information is added to the information bit sequence in order to facilitate error correction in the decoder. For example, in a systematic channel encoding scheme the redundant information is added to the information bit sequence as additional, inserted "coded" bits. In a non-systematic encoding scheme the outgoing bits are all coded bits, and there are no longer any "naked" information bits. The number of incoming bits (information bits) at the encoder is smaller than the number of outgoing bits (information bits plus inserted coded bits, or all coded bits). The ratio of incoming/outgoing bits is called the "code rate R" (typically $R=1/2$).

[0004] Recent improvements using the "Turbo Principle" have shown that, in digital communication systems involving a plurality of users in wireless communication with a receiver, an improvement in the quality of the decoded signal can be achieved by applying iterative decoding steps to the received data. In particular, "Iterative Equalization and Decoding in Mobile Communication Systems" by Bauch, Khorram and Hagenauer, EPMCC'97, pp 307-312, October 1997, Bonn, Germany, discusses the application of the Turbo principle to iterative decoding of coded data transmitted over a mobile radio channel.

[0005] In order to be suitable for iterative decoding, a transmitted signal must be encoded by at least two concatenated codes, either serially or parallelly concatenated.

[0006] Fig. 1 shows a serially concatenated coding scheme: the transmission is done on a block-by-block basis. The binary signal from the digital source is encoded firstly by an outer encoder and is then passed through an interleaver, which changes the order of the incoming bit symbols to make the signal appear more random to the following processing stages. After the interleaver, the signal is encoded a second time by an 'inner encoder'. Correspondingly, at the receiver the signal is first decoded by the inner decoder in a first decod-

ing step, deinterleaved, and decoded by the outer decoder in a second decoding step. From the outer decoder soft decision values are fed back as additional *a priori* input to the inner decoder. The soft decision values provide information on the reliability of the hard decision values. In a first iteration the decoding step is repeated and the soft decision values are used as input values for the first and second decoder.

[0007] The iterative decoding of a particular transmitted sequence is stopped with an arbitrary termination criterion, e.g. after a fixed number of iterations, or until a certain bit error rate is reached. It should be noted that the *a priori* soft value input to the inner decoder is set to zero for the very first decoding of the transmitted bit sequence ('0th iteration').

[0008] The inner and outer binary codes can be of any type: systematic, or non-systematic, block or convolutional codes. Simple mapping (e.g. antipodal or binary phase shift keying) is performed in the transmitter (after the inner encoder) and simple demapping is performed in the receiver (after the inner decoder) although for clarity this is not shown in Figure 1. Likewise, Figure 1 illustrates a single user scenario, although application of appropriate multiplexing provides a suitable multi user system.

[0009] At the receiver the two decoders are soft-in/soft-out decoders (SISO-decoder). A soft value represents the reliability on the bit decision of the respective bit symbol (whether 0 or 1 was sent). A soft-in decoder accepts soft reliability values for the incoming bit symbols. A soft-out decoder provides soft reliability output values on the outgoing bit symbols. The soft-out reliability values are usually more accurate than the soft-in reliability values since they are improved during the decoding process, based on the redundant information added with each encoding step at the transmitter. The best performance is achieved by a SISO-decoder which provides the *A Posteriori* Probability calculator (APP), tailored to the respective channel code. Several faster, but sub-optimal algorithms exist, e.g. the SOVA (soft output Viterbi algorithm).

[0010] In multilevel modulation, M bits (bit symbols) are grouped together at the transmitter to form one 'mapped symbol' (also briefly referred to as 'symbol'). This symbol can be mapped onto a real or a complex signal space (i.e. real axis, or complex plane). The mapping operation simply associates the unmapped symbol (M bits, value from 0, ..., $2^M - 1$) with a discrete amplitude level for Pulse Amplitude Modulation (PAM), a discrete phase level for Phase Shift Keying (PSK), any discrete signal point in the complex plane for quadrature Amplitude Modulation (QAM) or any combination of PAM, QAM, PSK. The mapping can be of any type.

[0011] At the receiver the incoming symbols are noise affected. The hard decision demapping operation associates the incoming symbol with the closest signal point in the signal space (signal point with minimum Euclidian distance in real or complex signal space) and

takes for example the respective Gray-encoded code-word as the hard decision values (0,1) for the M bits per mapped symbol.

[0012] However, if multilevel modulation is used in conjunction with channel coding and soft channel decoding (i.e. a soft input decoder) the demapping operation preferably calculates soft reliability values as inputs to the channel decoder. For simplicity, the term "multilevel modulation" is used when referring to PAM, PSK or QAM modulation, meaning "multi-amplitude level" for PAM, "multi phase level" for PSK, and "multi signal points" for QAM.

[0013] In one prior proposal, apparatus for iteratively decoding a signal has a demapper which has a first input for receiving the signal and an output for generating a demapped signal; and a decoder which has an input for receiving the demapped signal and an output for generating a decoded signal, the demapper having a second input for receiving the decoded signal.

[0014] Each user in a mobile communication system may have a different Quality of Service (QoS) requirement, i.e. different BER and latency constraints due to differing communication services. For example: voice communication has the lowest BER requirements (i.e. can tolerate many bit errors) with the highest latency constraints (i.e. cannot tolerate long delays in two way conversation); visual communication has a higher BER requirement and high latency constraints; data communication (e.g. wireless Internet web-browsing) has the highest BER requirements and the lowest latency constraints. Each user communicates with the base station with a different signal quality (i.e. SNR), multipath propagation and fading due to differing distance from the base station, propagation environment and, if mobile, speed.

[0015] The mapping operation itself does not add redundancy (in contrast to the inner encoder in classic serially concatenated encoding schemes) to the signal, but links bits together by grouping several bit symbols to form one mapped symbol.

[0016] The demapper is a soft demapping device that has been modified in order to accept a *priori* information obtained from the decoder. The decoder is a channel decoder and can be any SISO-decoder (optimal APP, or other sub-optimal algorithm, e.g. SOVA). The iterative demapping and decoding can thus be regarded as a serially concatenated iterative decoding scheme whereby the inner decoder is replaced by the soft demapping device. The iterative demapping and decoding is stopped by an arbitrary termination criterion (e.g. after a fixed number of iterations, or when a certain bit error rate is reached).

[0017] Mutual information is a parameter taken from information theory, see "The Elements of Information Theory" by T.M. Cover and J.A. Thomas. It specifies the maximal possible throughput for a given communication channel condition.

[0018] If we regard the mapping/demapping as part

of the communication channel we can describe the maximal throughput by means of unconditional bit-wise mutual information I_0 of the particular mapping.

[0019] The mean unconditional bit-wise mutual information I_0 for no other bits of the mapping known is defined as

$$I_0 = I(X_k; Y) = \frac{1}{M} \cdot \sum_{k=0}^{M-1} I(X_k; Y)$$

where

M is the number of bits of the mapping (e.g. for a 16 QAM-mapping $M=4$, since $2^4 = 16$)

X_k is the k^{th} bit of the mapping, with $k=0 \dots M-1$; input variable to the communication channel

Y is the output variable of the communication channel.

$I(X_k; Y)$ and thus I_0 are dependent on the applied mapping and the signal-to-noise ratio (typically given by E_b/N_0), and can be calculated according to methods given in "Elements of Information Theory" referred to above.

[0020] We have found by simulation, that the best mapping in an iterative demapping (IDEM) system depends on the E_b/N_0 -region of interest and on the number of iterations N_{bit} that can be performed at the receiver (see Fig. 2). As explained before, this best mapping is determined by its unconditional bit-wise mutual information I_0 . Hence, for the optimal mapping we have

$$I_{0,\text{opt}} = \text{function of } (E_b/N_0, N_{\text{bit}}).$$

[0021] In accordance with one aspect of the invention, there is provided a digital transmission system, comprising:

a coder for coding an input signal with an error checking or error correcting code, an interleaver for interleaving the bits of the coded signal, a mapper for mapping the interleaved coded signal into a multilevel signal, and a transmitter for sending the multilevel signal over a noisy channel in a transmission medium;

a receiver for receiving the multilevel signal distorted by noise (noisy multilevel signal), a demapper having a first input for receiving the noisy multilevel signal and operative to provide a soft demapped signal at an output, a deinterleaver, for deinterleaving the demapped signal, a decoder to decode the deinterleaved demapped signal to pro-

vide a soft decoded signal, an interleaver to interleave the decoded signal, the demapper having a second input for receiving the interleaved decoded signal and being operative to recalculate the demapped signal iteratively using the noisy signal and the soft decoded signal, characterised by:

a mapping store for storing two different mappings having different mutual informations;

means for deriving an indication of the channel conditions;

means for selecting the mappings in a mixing ratio dependent on the derived indication of channel conditions, the mapper and demapper being operatively responsive to the selected mapping.

[0022] In accordance with another aspect of the invention, there is provided a digital transmission system, comprising:

a coder for coding an input signal with an error checking or error correcting code, an interleaver for interleaving the bits of the coded signal, a mapper for mapping the interleaved coded signal into a multilevel signal, and a transmitter for sending the multilevel signal over a noisy channel in a transmission medium;

a receiver for receiving the multilevel signal distorted by noise (noisy multilevel signal), a demapper having a first input for receiving the noisy multilevel signal and operative to provide a soft demapped signal at an output, a deinterleaver, for deinterleaving the demapped signal, a decoder to decode the deinterleaved demapped signal to provide a soft decoded signal, an interleaver to interleave the decoded signal, the demapper having a second input for receiving the interleaved decoded signal and being operative to recalculate the demapped signal iteratively using the noisy signal and the soft decoded signal, characterised by:

a mapping store for storing two different mappings having different mutual informations;

means for determining the number of iterations to be carried out from a plurality of possible numbers of iteration; and

means for selecting the mappings in a mixing ratio dependent on the number of iterations as determined, the mapper and demapper being operatively responsive to the selected mapping.

[0023] In accordance with yet another aspect of the invention there is provided a digital transmission system, comprising:

a coder for coding an input signal with an error checking or error correcting code, an interleaver for interleaving the bits of the coded signal, a mapper

for mapping the interleaved coded signal into a multilevel signal, and a transmitter for sending the multilevel signal over a noisy channel in a transmission medium;

a receiver for receiving the multilevel signal distorted by noise (noisy multilevel signal), a demapper having a first input for receiving the noisy multilevel signal and operative to provide a soft demapped signal at an output, a deinterleaver, for deinterleaving the demapped signal, a decoder to decode the deinterleaved demapped signal to provide a soft decoded signal, an interleaver to interleave the decoded signal, the demapper having a second input for receiving the interleaved decoded signal and being operative to recalculate the demapped signal iteratively using the noisy signal and the soft decoded signal, characterised by:

a mapping store for storing two different mappings having different mutual informations;

means for deriving an indication of the channel conditions;

means for determining the number of iterations to be carried out from a plurality of possible numbers of iterations; and

means for selecting the mappings in a mixing ratio dependent on the derived indication of channel conditions and on the number of iterations as determined, the mapper and demapper being operatively responsive to the selected mapping.

[0024] Preferably the indication of the channel conditions is an estimate of the signal to noise ratio.

[0025] The soft output signals of the demapper and the decoder are preferably likelihood ratios, most preferably log-likelihood ratios.

[0026] Each mapping is preferably stored with an indication of its mutual information, and the system includes a mutual information store storing an approximation of the optimum mutual information with a plurality of respective combinations of channel conditions and number of iterations, the means for selecting being operative to identify an optimum mutual information from the number of iterations as determined and the nearest stored channel conditions to the derived indication, and to select the mapping in a mixing ratio to give a mutual information for the mix approximately equal to the optimum.

[0027] The selecting means is preferably operative to calculate the mixing ratio q according to the formula:

$$q = \frac{I_{o,max} - I_{o,opt}}{I_{o,opt} - I_{o,min}}$$

where

$I_{o,max}$ is the larger mutual information of the stored

mappings.

$I_{0,min}$ is the smaller mutual information of the stored mappings, and

$I_{0,opt}$ is the desired optimum mutual information.

[0028] The invention also extends to a method of transmitting a digital signal, comprising:

coding an input signal with an error checking or error correcting code, interleaving the bits of the coded signal, mapping the interleaved coded signal into a multilevel signal, sending the multilevel signal over a noisy channel in a transmission medium;

receiving the multilevel signal distorted by noise (noisy multilevel signal), demapping the noisy multilevel signal to provide a soft demapped signal, deinterleaving the demapped signal, decoding the deinterleaved demapped signal to provide a soft decoded signal, interleaving the decoded signal, recalculating the demapped signal iteratively using the noisy signal and the soft decoded signal,

characterised by:

storing a plurality of different mappings;

deriving an indication of the channel conditions; and

selecting the optimum mapping of the plurality dependent on the derived indication of channel conditions, the mapping and demapping being operatively responsive to the selected mapping and demapping.

[0029] The invention further extends to a method of transmitting a digital signal, comprising:

coding an input signal with an error checking or error correcting code, interleaving the bits of the coded signal, mapping the interleaved coded signal into a multilevel signal, sending the multilevel signal over a noisy channel in a transmission medium;

receiving the multilevel signal distorted by noise (noisy multilevel signal), demapping the noisy multilevel signal to provide a soft demapped signal, deinterleaving the demapped signal, decoding the deinterleaved demapped signal to provide a soft decoded signal, interleaving the decoded signal, recalculating the demapped signal iteratively using the noisy signal and the soft decoded signal, characterised by:

storing a plurality of different mappings;

determining the number of iterations to be carried out from a plurality of possible numbers of iteration; and

selecting the optimum mapping of the plurality, dependent on the number of iterations as determined, the mapping and demapping being operatively responsive to the selected mapping.

[0030] The invention yet further extends to a

method of transmitting a digital signal, comprising:

coding an input signal with an error checking or error correcting code, interleaving the bits of the coded signal, mapping the interleaved coded signal into a multilevel signal, sending the multilevel signal over a noisy channel in a transmission medium;

receiving the multilevel signal distorted by noise (noisy multilevel signal), demapping the noisy multilevel signal to provide a soft demapped signal, deinterleaving the demapped signal, decoding the deinterleaved demapped signal to provide a soft decoded signal, interleaving the decoded signal, recalculating the demapped signal iteratively using the noisy signal and the soft decoded signal,

characterised by:

storing a plurality of different mappings;

deriving an indication of the channel conditions; means for determining the number of iterations to be carried out from a plurality of possible numbers of iterations; and

selecting the optimum mapping of the plurality dependent on the derived indication of channel conditions and on the number of iterations as determined, the mapping and demapping being operatively responsive to the selected mapping.

[0031] An embodiment of the invention will now be described, by way of example, with reference to the accompanying drawings, in which:

Figure 1 is a block diagram of a prior art iterative decoding system;

Figure 2 shows achieved bit error rate plotted against mean unconditional bit-wise mutual information I_0 at one value of E_b/N_0 in the channel;

Figure 3 shows iteration plotted against mean unconditional bitwise mutual information I_0 for the best bit error rate estimated from Figure 2;

Figure 4 shows a block diagram of a transmission system embodying the invention; and

Figures 5 to 7 show different mappings.

[0032] Referring to Figure 4, the transmitter, a binary random signal from a source 2 is convolutionally encoded by coder 4 and fed to an interleaver 6 which interleaves the bit symbols. (Any channel code can be used, non-systematic convolutional codes are used merely as an example). After the interleaver, M bits are grouped together and mapped onto a complex signal constellation by a mapper 8, according to the applied modulation scheme

[0033] In the channel, the symbols become distorted by additive noise and any other noise form.

[0034] At the receiver the channel symbols are demapped and ungrouped by a log-likelihood ratio calculation in a demapper 10 for each of the M bits per symbol. The log-likelihood ratio values (soft values) are

deinterleaved in a deinterleaver 12 and put into a coder 14 in the form of an *A Posteriori Probability* calculator (APP). (Any other SISO-decoder may be used). After the decoding the estimates on the transmitted information bits are available at the output of a hard decision device 16 by taking the sign of the APP-soft output values for the information bits.

[0035] In the iterative demapping/decoding path the extrinsic information is passed through a bit interleaver 22 and fed back as *a priori* knowledge to the soft demapping device. The extrinsic information is the difference between the soft input and the soft output value at the decoder, and depicts the new, statistically independent information (at least for the first iteration) gained by the decoding process.

[0036] The complex channel symbol z at the receiver can be considered as a matched filter output. It carries M encoded bits. Since the SISO-decoder has soft input processing, the demapping device extracts a soft value for each bit x_0, \dots, x_{N-1} for further decoding in the SISO-decoder. This soft value for each of the M bits per symbol is the log-likelihood ratio (L-value) of the respective bit conditioned on the matched filter output z . The absolute value of the L-value denotes the reliability of the bit decision.

[0037] The full term of the L-value calculation for bit x_k consists of an additive '*a priori*' L-value for bit x_k and a fractional term in which the *a priori* L-values of the remaining, bits $x_{j,j=0 \dots N-1, j \neq k}$ are included.

[0038] The *a priori* L-values of bits x_0, \dots, x_{N-1} are provided by the SISO-decoder as inputs to the soft demapping device.

[0039] Simulations show that the best performance of iterative soft demapping and decoding is achieved if the additive *a priori* L-value for bit x_k is left out of the full term of the L-value for bit x_k , and if the *a priori* L-values of the remaining bits $x_{j,j=0 \dots N-1, j \neq k}$ are considered in the calculation of the L-value for bit x_k . This is indicated in Fig. 2 by the subtraction after the demapping device: the *a priori* values coming from the SISO-decoder are subtracted from the output of the log-likelihood ration calculation of the respective bit in the demapping device. The information that is fed to the deinterleaver can thus be regarded as the 'extrinsic information' of the demapping device (in contrast to the extrinsic information from the SISO-decoder).

[0040] Note the L-value calculation implies both, soft demapping and ungrouping of the M bits per symbol (not two separate operations, as Fig.2 might suggest).

[0041] Iterative soft demapping and decoding reduces the bit error rate of conventional multilevel modulation schemes that use plain channel coding. Many modern digital communications systems, with simple channel coding and multilevel modulation, may be improved by altering the receiver circuitry to include a soft demapping device that accepts *a priori* information, and a SISO-decoder as channel decoder.

[0042] It is applicable to multilevel modulation schemes with M bits per symbol, whereby $M > 1$ for PAM, PSK and QAM, whereby for PSK and QAM with $M=2$ Anti-Gray-mapping has to be applied.

[0043] It is important to note that the interleaver is a bit symbol interleaver, which interleaves the symbol on the bit level. Providing there is at least one bit symbol interleaver between encoder and mapper, other systems that apply both bit symbol and 'n bit' symbol interleavers in a serial concatenation between encoder and symbol mapper may be employed.

[0044] A signal to noise ratio estimator 18 estimates the signal to noise ratio in the channel signal. A controller 20 determines, amongst other things, how many iterations can be used for the channel. The number of iterations could change for a number of reasons, e.g. in order to reduce power consumption, the receiver could reduce the number of iterations, or in a multi-user receiver computing resources available for iterative decoding may be shifted from one user to another higher priority user.

[0045] We have found by experimental simulation, that the best mapping in an IDEM system depends on the E_b/N_0 -region of interest and on the number of iterations N_{bit} that can be performed at the receiver. Figure 2 shows the bit error ratio plotted against mean unconditional bitwise mutual information for different numbers of iteration, all at one signal to noise ratio. Different plots are obtained at different signal to noise ratios. As can be seen given the signal to noise conditions in the channel and the number of iterations which the receiver will perform, there is a minimum bit error rate in the iteratively decoded signal which occurs at a particular mean unconditional bit-wise mutual information I_0 . Estimates of the iterations and I_0 at which the minima occur, for that E_b/N_0 , are plotted in Figure 3. Similar results are obtained for a range of different signal to noise ratios E_b/N_0 . The results may be stored in an I_0 look-up table.

[0046] Two mappings are generated and their mean unconditional bitwise mutual informations I_0 are calculated. Examples at the extremes are shown in Figure 5 and 7. The mapping shown in Figures 5 is a Gray mapping and has $I_0 = 0.54$ bit per complex dimension at $E_b/N_0 = 3\text{dB}$ and code rate 1:2. Figure 7 is another mapping for 16QAM and has $I_0 = 0.36$ bit per complex dimension at $E_b/N_0 = 3\text{dB}$.

[0047] For a given combination of channel conditions and number of iterations, the channel conditions for which results have been obtained and which are nearest to the estimated conditions are used in combination with the number of iterations, to identify the best mapping. The mappings are stored in a mapping look up table with the respective I_0 . Having found the required I_0 from the I_0 look-up table, the desired mapping is generated by mixing the mappings in a mixing ratio q according to the formula:

$$q = \frac{I_{o,max} - I_{o,opt}}{I_{o,opt} - I_{o,min}}$$

where

$I_{o,max}$ is the larger mutual information of the stored mappings,

$I_{o,min}$ is the smaller mutual information of the stored mappings, and

$I_{o,opt}$ is the desired optimum mutual information.

[0048] Thus a period of $c = a + b$ consecutively transmitted M-bit symbol, contains

$$a = \frac{q}{1+q} \cdot c$$

of the mappings with $I_{o,min}$, and

$$b = \frac{1}{1+q} \cdot c$$

of the mappings with $I_{o,max}$.

[0049] The period c is arbitrary but, as an example, could be about 50.

[0050] The mixing ratio can be determined at the receiver and transmitted to the transmitter. Alternatively the number of iterations to be used and the estimated channel conditions can be transmitted to the transmitter where the mixing ratio is determined and then transmitted to the receiver.

[0051] It is also possible to mix more than two mappings. Figure 6 shows a mapping for 16QAM and has an intermediate $I_o = 0.23$ bit per complex dimension at $E_b/N_o = 3$ dB.

Claims

1. A digital transmission system, comprising:

a coder for coding an input signal with an error checking or error correcting code, an interleaver for interleaving the bits of the coded signal, a mapper for mapping the interleaved coded signal into a multilevel signal, and a transmitter for sending the multilevel signal over a noisy channel in a transmission medium; a receiver for receiving the multilevel signal distorted by noise (noisy multilevel signal), a demapper having a first input for receiving the noisy multilevel signal and operative to provide a soft demapped signal at an output, a deinterleaver, for deinterleaving the demapped signal, a decoder to decode the deinterleaved demapped signal to provide a soft decoded signal,

an interleaver to interleave the decoded signal, the demapper having a second input for receiving the interleaved decoded signal and being operative to recalculate the demapped signal iteratively using the noisy signal and the soft decoded signal, characterised by:

a mapping store for storing two different mappings having different mutual informations;

means for deriving an indication of the channel conditions;

means for selecting the mappings in a mixing ratio dependent on the derived indication of channel conditions, the mapper and demapper being operatively responsive to the selected mapping.

2. A digital transmission system, comprising:

a coder for coding an input signal with an error checking or error correcting code, an interleaver for interleaving the bits of the coded signal, a mapper for mapping the interleaved coded signal into a multilevel signal, and a transmitter for sending the multilevel signal over a noisy channel in a transmission medium; a receiver for receiving the multilevel signal distorted by noise (noisy multilevel signal), a demapper having a first input for receiving the noisy multilevel signal and operative to provide a soft demapped signal at an output, a deinterleaver, for deinterleaving the demapped signal, a decoder to decode the deinterleaved demapped signal to provide a soft decoded signal; an interleaver to interleave the decoded signal; the demapper having a second input for receiving the interleaved decoded signal and being operative to recalculate the demapped signal iteratively using the noisy signal and the soft decoded signal, characterised by:

a mapping store for storing two different mappings having different mutual informations;

means for determining the number of iterations to be carried out from a plurality of possible numbers of iteration; and

means for selecting the mappings in a mixing ratio dependent on the number of iterations as determined, the mapper and demapper being operatively responsive to the selected mapping.

3. A digital transmission system, comprising:

a coder for coding an input signal with an error checking or error correcting code, an interleaver for interleaving the bits of the coded signal,

- a mapper for mapping the interleaved coded signal into a multilevel signal, and a transmitter for sending the multilevel signal over a noisy channel in a transmission medium;
- a receiver for receiving the multilevel signal distorted by noise (noisy multilevel signal), a demapper having a first input for receiving the noisy multilevel signal and operative to provide a soft demapped signal at an output, a deinterleaver, for deinterleaving the demapped signal, a decoder to decode the deinterleaved demapped signal to provide a soft decoded signal, an interleaver to interleave the decoded signal, the demapper having a second input for receiving the interleaved decoded signal and being operative to recalculate the demapped signal iteratively using the noisy signal and the soft decoded signal, characterised by:
- a mapping store for storing two different mappings having different mutual informations;
- means for deriving an indication of the channel conditions;
- means for determining the number of iterations to be carried out from a plurality of possible numbers of iterations; and
- means for selecting the mappings in a mixing ratio dependent on the derived indication of channel conditions and on the number of iterations as determined, the mapper and demapper being operatively responsive to the selected mapping.
4. A digital transmission system, as claimed in claim 1 or 3, wherein the indication of the channel conditions is an estimate of the signal to noise ratio.
 5. A digital transmission system as claimed in any preceding claim, in which the soft output signals of the demapper and the decoder are likelihood ratios.
 6. A digital transmission system as claimed in any preceding claim, in which the soft output signals of the demapper and the decoder are log-likelihood ratios.
 7. A digital transmission system, as claimed in claim 3 or in any preceding claim dependent on claim 3, wherein each mapping is stored with an indication of its mutual information, and including a mutual information store storing an approximation of the optimum mutual information with a plurality of respective combinations of channel conditions and number of iterations, the means for selecting being operative to identify an optimum mutual information from the number of iterations as determined and the nearest stored channel conditions to the derived indication, and to select the mapping in a mixing ratio to give a mutual information for the mix

approximately equal to the optimum.

8. A digital transmission system as claimed in claim 7, wherein the selecting means is operative to calculate the mixing ratio q according to the formula:

$$q = \frac{I_{o,max} - I_{o,opt}}{I_{o,opt} - I_{o,min}}$$

where

$I_{o,max}$ is the larger mutual information of the stored mappings,
 $I_{o,min}$ is the smaller mutual information of the stored mappings, and
 $I_{o,opt}$ is the desired optimum mutual information.

9. A method of transmitting a digital signal, comprising:

coding an input signal with an error checking or error correcting code, interleaving the bits of the coded signal, mapping the interleaved coded signal into a multilevel signal, sending the multilevel signal over a noisy channel in a transmission medium;

receiving the multilevel signal distorted by noise (noisy multilevel signal), demapping the noisy multilevel signal to provide a soft demapped signal, deinterleaving the demapped signal, decoding the deinterleaved demapped signal to provide a soft decoded signal, interleaving the decoded signal, recalculating the demapped signal iteratively using the noisy signal and the soft decoded signal,

characterised by:

storing a plurality of different mappings;

deriving an indication of the channel conditions; and

selecting the optimum mapping of the plurality dependent on the derived indication of channel conditions, the mapping and demapping being operatively responsive to the selected mapping and demapping.

10. A method of transmitting a digital signal, comprising:

coding an input signal with an error checking or error correcting code, interleaving the bits of the coded signal, mapping the interleaved coded signal into a multilevel signal, sending the multilevel signal over a noisy channel in a transmission medium;

receiving the multilevel signal distorted by noise (noisy multilevel signal), demapping the

noisy multilevel signal to provide a soft demapped signal at an output, deinterleaving the demapped signal, decoding the deinterleaved demapped signal to provide a soft decoded signal, interleaving the decoded signal, recalculating the demapped signal iteratively using the noisy signal and the soft decoded signal, characterised by:

storing a plurality of different mappings;
determining the number of iterations to be carried out from a plurality of possible numbers of iteration; and
selecting the optimum mapping of the plurality, dependent on the number of iterations as determined, the mapping and demapping being operatively responsive to the selected mapping.

11. A method of transmitting a digital signal, comprising:

coding an input signal with an error checking or error correcting code, interleaving the bits of the coded signal, mapping the interleaved coded signal into a multilevel signal, sending the multilevel signal over a noisy channel in a transmission medium;

receiving the multilevel signal distorted by noise (noisy multilevel signal), demapping the noisy multilevel signal to provide a soft demapped signal, deinterleaving the demapped signal, decoding the deinterleaved demapped signal to provide a soft decoded signal, interleaving the decoded signal, recalculating the demapped signal iteratively using the noisy signal and the soft decoded signal, characterised by:

storing a plurality of different mappings;
deriving an indication of the channel conditions;
determining the number of iterations to be carried out from a plurality of possible numbers of iterations; and
selecting the optimum mapping of the plurality dependent on the derived indication of channel conditions and on the number of iterations as determined, the mapping and demapping being operatively responsive to the selected mapping.

12. A method of transmitting a digital signal, as claimed in claim 9 or 11, wherein the indication of the channel conditions is the signal to noise ratio.

13. A method of transmitting a digital signal as claimed in any preceding claim, in which the soft demapped and soft decoded signals are likelihood ratios.

14. A method of transmitting a digital signal as claimed in any preceding claim, in which the soft demapped and soft decoded signals are log-likelihood ratios.

15. A method of transmitting a digital signal, as claimed in claim 11 or in any preceding claim dependent on claim 11, wherein each mapping is stored with an indication of its mutual information, and including storing an approximation of the optimum mutual information with a plurality of respective combinations of channel conditions and number of iterations, the step of selecting a mapping being operative to identify an optimum mutual information from the number of iterations as determined and the nearest stored channel conditions to the derived indication,

and selecting the mappings in a mixing ratio to give a mutual information for the mix approximately equal to the optimum.

16. A method as claimed in claim 15, wherein the mixing ratio q is calculated according to the formula:

$$q = \frac{I_{o,max} - I_{o,opt}}{I_{o,opt} - I_{o,min}}$$

where

$I_{o,max}$ is the larger mutual information of the stored mappings,

$I_{o,min}$ is the smaller mutual information of the stored mappings, and

$I_{o,opt}$ is the desired optimum mutual information.

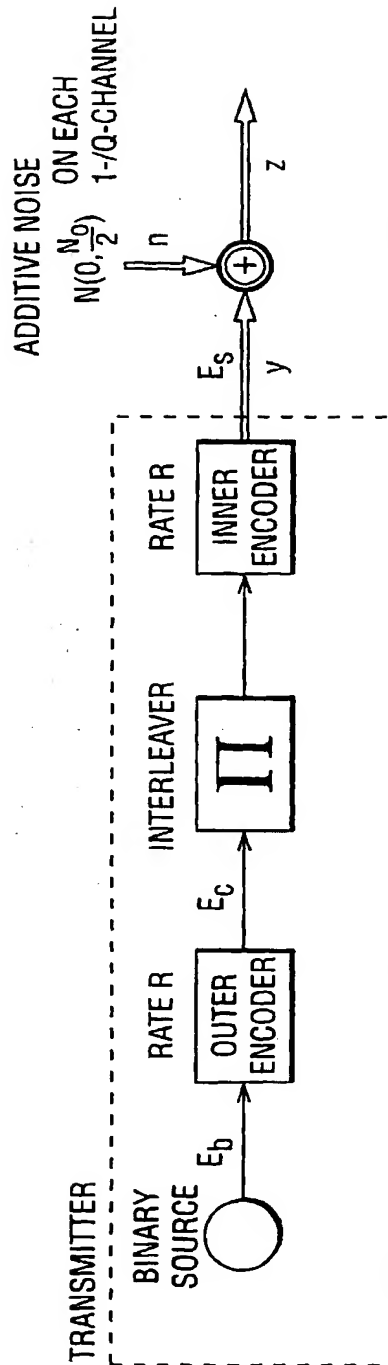
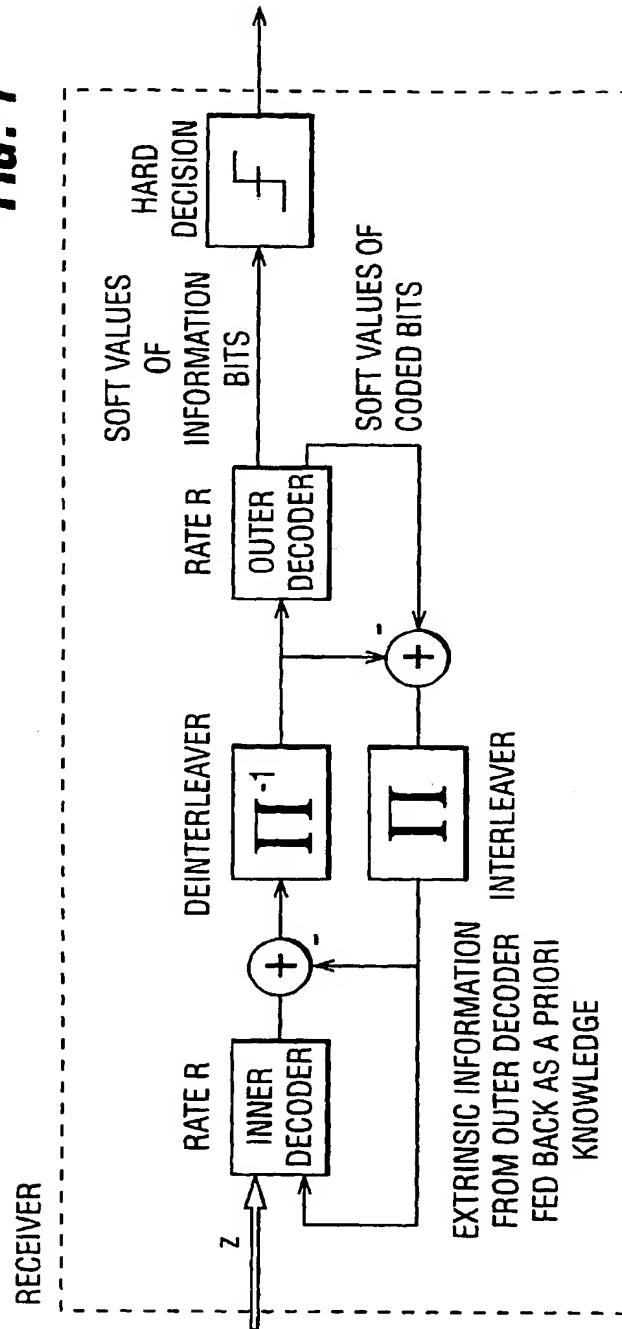


FIG. 1



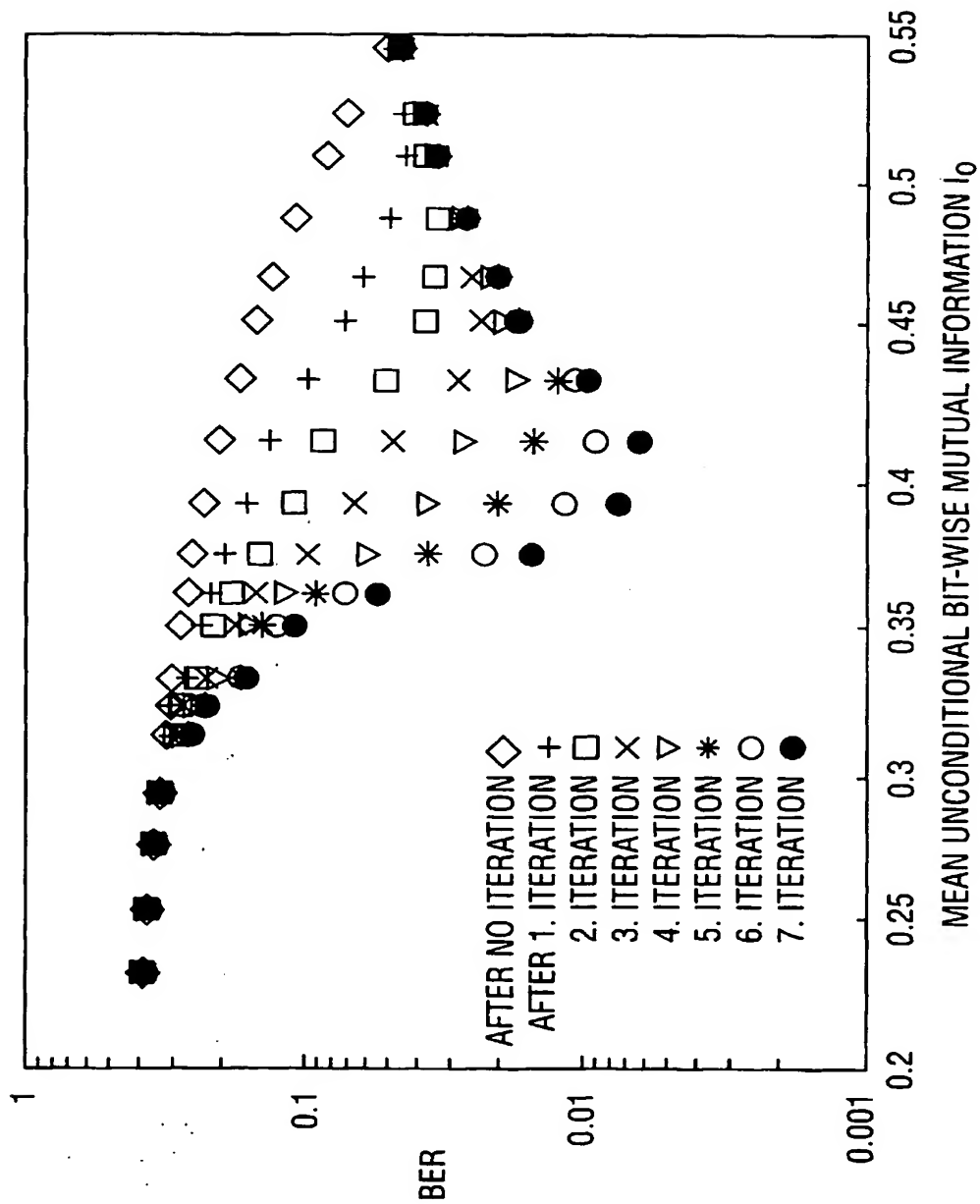


FIG. 2

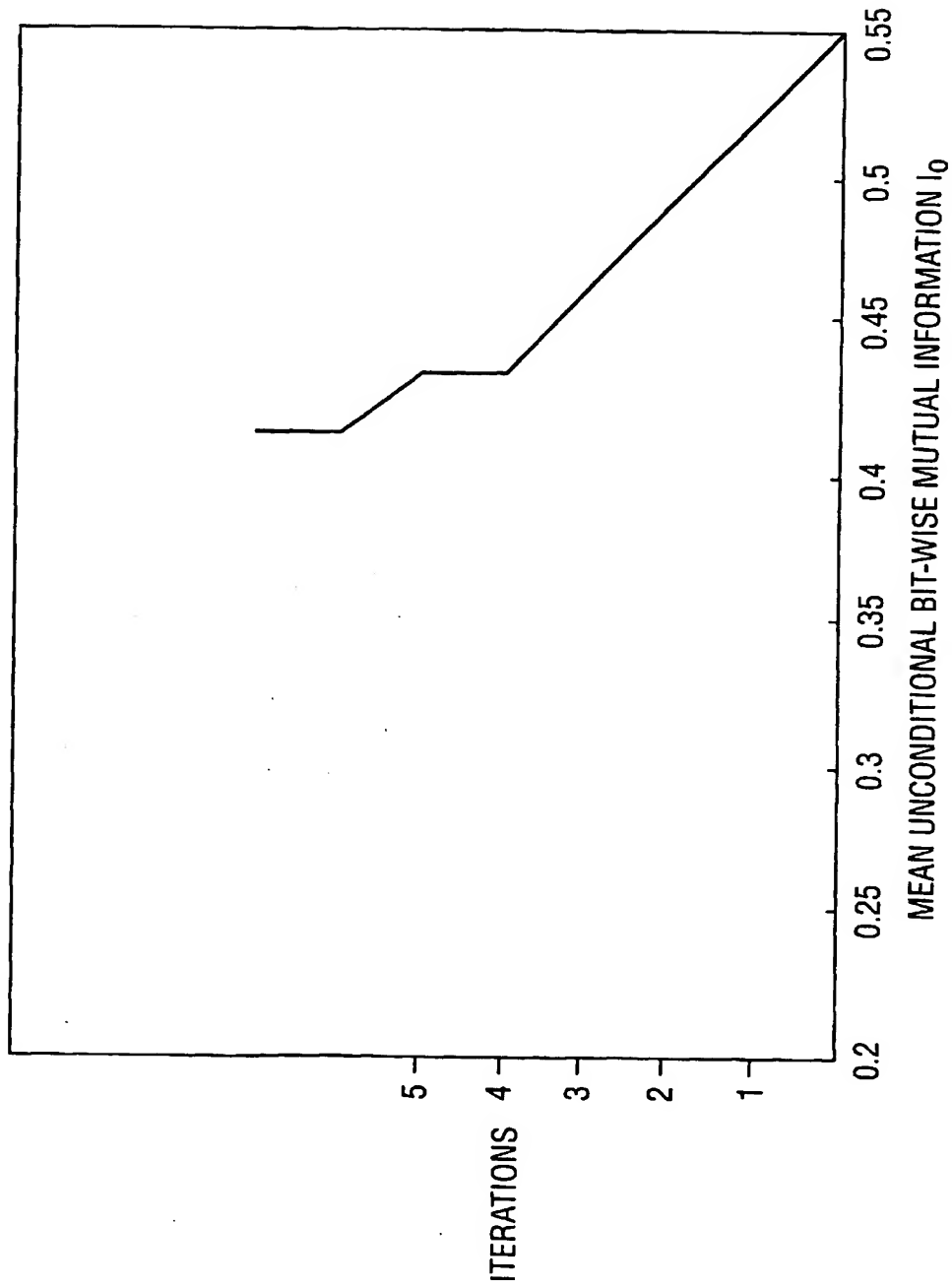
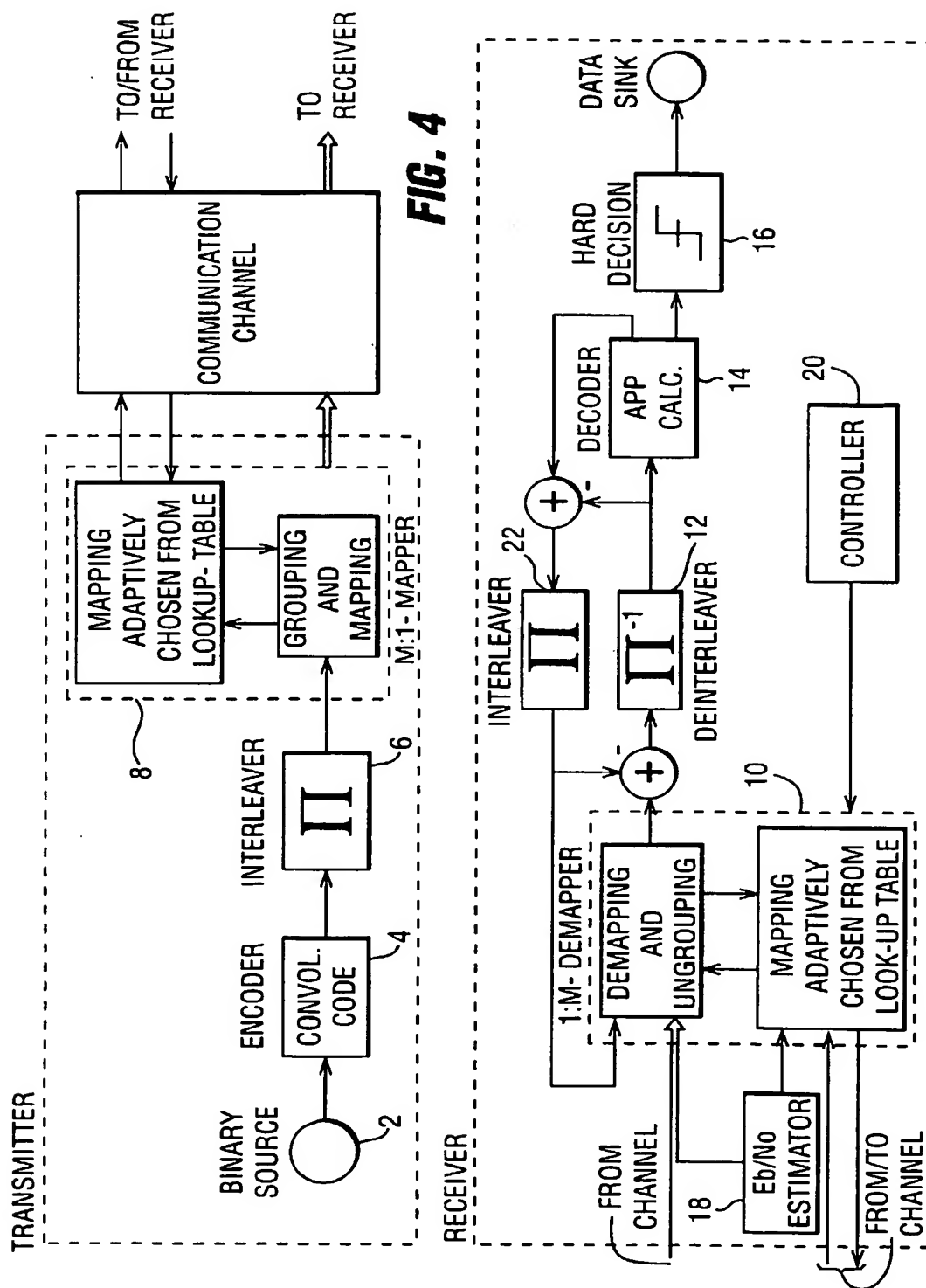


FIG. 3



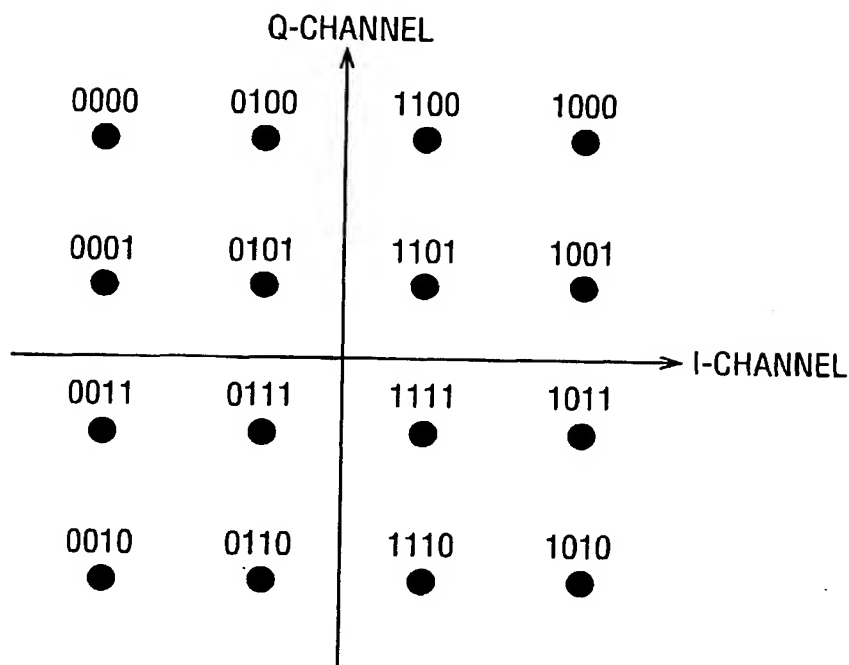


FIG. 5

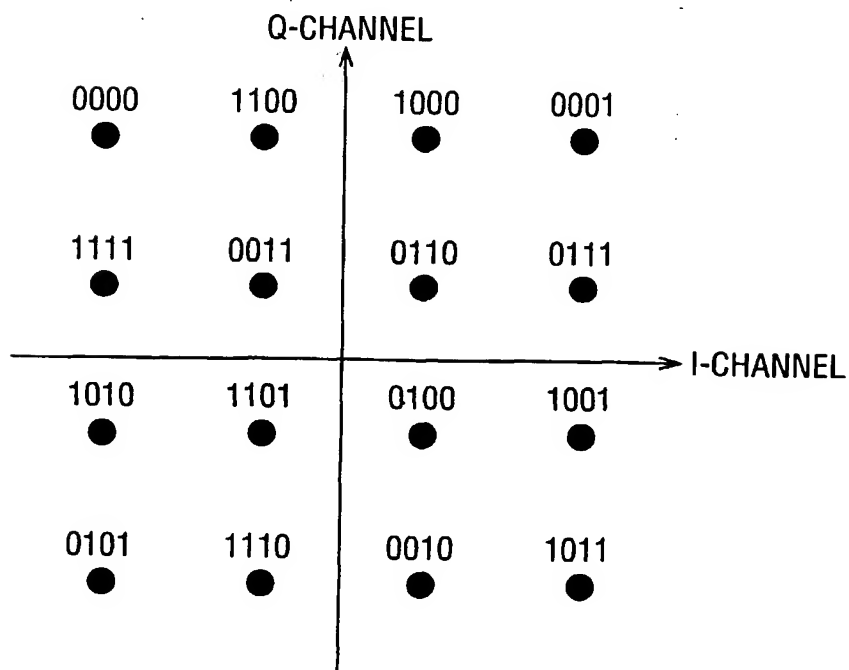


FIG. 6

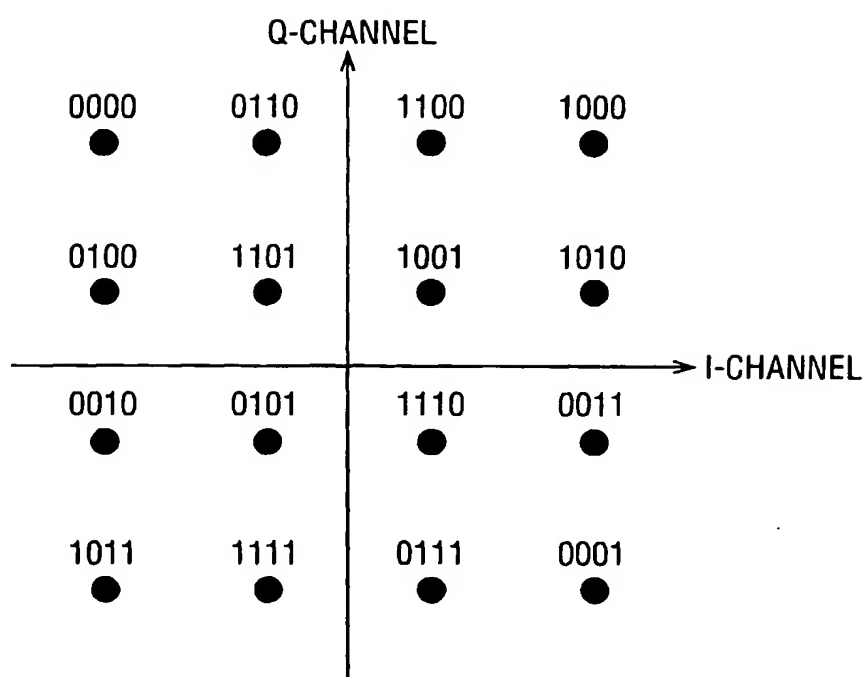


FIG. 7



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EUROPEAN SEARCH REPORT

Application Number
EP 98 30 8893

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A	FAZEL K ET AL: "COMBINED MULTILEVEL TURBO-CODE WITH 8PSK MODULATION" GLOBECOM '95. IEEE GLOBAL TELECOMMUNICATIONS CONFERENCE, SINGAPORE, vol. 1, 14 - 16 November 1995, pages 649-653, XP000621563 INSTITUTE OF ELECTRICAL AND ELECTRONICS ENGINEERS * section 2.2 * * section 3.2 * * section 5 * * figure 3 *	1-3,9-11	
A	WO 94 11955 A (PERICLE COMMUNICATIONS CO) 26 May 1994 * page 37, line 13 - line 15 * * page 37, line 21 - line 23 * * figure 1 *	1-3,9-11	TECHNICAL FIELDS SEARCHED (Int.Cl.6) H04L
The present search report has been drawn up for all claims			
Place of search THE HAGUE		Date of completion of the search 15 April 1999	Examiner Ghigliotti, L
CATEGORY OF CITED DOCUMENTS X : particularly relevant if taken alone Y : particularly relevant if combined with another document of the same category A : technological background O : non-written disclosure P : intermediate document T : theory or principle underlying the invention E : earlier patent document, but published on, or after the filing date D : document cited in the application L : document cited for other reasons & : member of the same patent family, corresponding document			

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Application Number
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			TECHNICAL FIELDS SEARCHED (Int.Cl.6)
The present search report has been drawn up for all claims			
Place of search THE HAGUE		Date of completion of the search 15 April 1999	Examiner Ghigliotti, L
<p>CATEGORY OF CITED DOCUMENTS</p> <p>X : particularly relevant if taken alone</p> <p>Y : particularly relevant if combined with another document of the same category</p> <p>A : technological background</p> <p>O : non-written disclosure</p> <p>P : intermediate document</p>		<p>T : theory or principle underlying the invention</p> <p>E : earlier patent document, but published on, or after the filing date</p> <p>O : document cited in the application</p> <p>L : document cited for other reasons</p> <p>& : member of the same patent family, corresponding document</p>	

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